Principles of Wireless Sensor Networks

https://www.kth.se/social/course/EL2745/

Lecture 13

Wireless Sensor Network Control Systems 2

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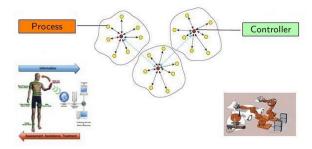
October 15, 2014

Course content

- Part 1
 - Lec 1: Introduction to WSNs
 - ► Lec 2: Introduction to Programming WSNs
- Part 2
 - ► Lec 3: Wireless Channel
 - ► Lec 4: Physical Layer
 - ► Lec 5: Medium Access Control Layer
 - ► Lec 6: Routing
- Part 3
 - ► Lec 7: Distributed Detection
 - ► Lec 8: Static Distributed Estimation
 - ► Lec 9: Dynamic Distributed Estimation
 - ► Lec 10: Positioning and Localization
 - ► Lec 11: Time Synchronization
- Part 4
 - ► Lec 12: Wireless Sensor Network Control Systems 1
 - ► Lec 13: Wireless Sensor Network Control Systems 2
 - Lec 14: Summary and Project Presentations

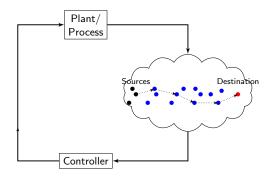
Previous lecture





How to model mathematically a closed loop control system?

Today's learning goals



- How stability is affected by delays introduced by the WSN?
- How stability is affected by packet losses introduced by the WSNs?
- How to design WSNCS?

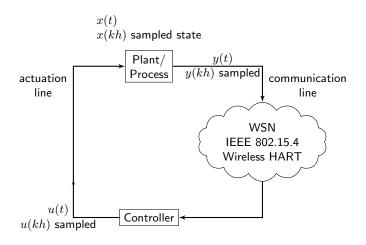
Outline

- Overview: WSNCS
- WSNCS with constant network delay
- WSNCS with random network delay
- WSNCS with rate constraint
- WSNCS with packet losses
- Design of WSNCS

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Overview: WSNCS



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WSNCS with constant network delay

Consider a linear model

$$\dot{x}(t) = Ax(t) + Bu(t)$$

$$y(t) = Cx(t)$$
(1)

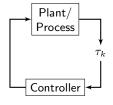
Assume that the controller takes a decision proportional to the state

$$u\left(t\right) = -Lx\left(t\right)$$

- L is chosen accordingly in order to achieve stability
- In general $u\left(t\right)=-L\hat{x}\left(t\right)$ where $\hat{x}\left(t\right)$ is an estimate of $x\left(t\right)$ based on $y\left(t\right)$
- Assume x(t) is estimated perfectly

WSNCS with constant network delay

Let au_k be the delay introduced by the network



Assume $0 \le \tau_k \le h$ where h the sampling time

When is the closed loop control system stable despite au_k and given $u\left(kh\right)=-Lx\left(kh\right)$?

Network Delays

Our model becomes:

$$\dot{x}(t) = Ax(t) + Bu(t - \tau_k), \quad t \in [kh, kh + h)$$

$$y(t) = Cx(t) + Du(t)$$
(2)

where u(t) is the input signal in the absence of delay, i.e. u(t)=u(kh) for $t\in [kh,kh+h).$

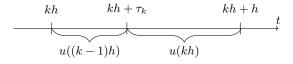
Integrating (2) yields:

$$x(kh+h) = \phi x(kh) + \Gamma_0(\tau_k)u(kh) + \Gamma_1(\tau_k)u(kh-h)$$
(3)

where
$$\phi=\int\limits_0^h e^{Ah}$$
 , $\Gamma_0(au_k)=\int\limits_0^{h- au_k} e^{As}dsB$ and $\Gamma_1(au_k)=\int\limits_{h- au_k}^h e^{As}dsB$

Let's see the detail in next slide...

WSNCS with constant network delay



From lecture 12, the solution of (1) is

$$x (kh + h) = e^{Ah} x (kh) + \int_{kh}^{kh+\tau_k} e^{A(t-\tau)} Bu (\tau) d\tau =$$

$$= e^{Ah} x (kh) + \int_{kh}^{kh+\tau_k} e^{A(t-\tau)} Bu ((k-1)h) d\tau + \int_{kh+\tau_k}^{kh+h} e^{A(t-\tau)} Bu (kh) d\tau =$$

$$= \phi x (kh) + \Gamma_0 (\tau_k) u (kh) + \Gamma_1 (\tau_k) u ((k-1)h)$$

where

$$\phi = e^{Ah} \qquad \Gamma_0 \left(\tau_k \right) = \int\limits_0^{h-\tau_k} e^{As} B ds \qquad \Gamma_1 \left(\tau_k \right) = \int\limits_{h-\tau_k}^h e^{As} B ds$$

Observe that Γ_0 and Γ_1 depend both on h and τ_k

Network Delays

Assume linear feedback:

$$u(x(kh)) = -Lx(kh), \quad K \in \mathbb{R}^{n \times n}$$

Then

$$x(kh+h) = \phi x(kh) - \Gamma_0(\tau_k) Lx(kh) - \Gamma_1(\tau_k) Lx((k-1)h)$$
(4)

Define the augmented state vector $z = \begin{bmatrix} x(kh) \\ u(kh-h) \end{bmatrix}$ and the matrix

$$\overline{\phi}(\tau_k) = \begin{bmatrix} \phi - \Gamma_0(\tau_k)K & \Gamma_1(\tau_k) \\ -K & 0 \end{bmatrix}.$$

Equation (4) is equivalent to

$$z(kh+h) = \overline{\phi}(\tau_k)z(kh) \tag{5}$$

Study the eigenvalues of $\overline{\phi}$ to determine the stability If the maximum eigenvalue of $\overline{\phi}$, $\rho\left(\overline{\phi}\right)<1\Rightarrow$ asymptotic stability

Example

Suppose that we are given a simple scalar system that is subject to a constant network delay au and governed by the following equation:

$$\dot{x}\left(t\right) = u\left(t\right)$$

Assume that the controller decision is u(t) = -Lx(t).

In order to study the stability of the delay system, the matrix $\overline{\phi}$ needs to be constructed. Thus, since A=0 and B=1,

$$\phi = e^0 = 1$$
 $\Gamma_0 = \int_0^{h-\tau} ds = h - \tau$ $\Gamma_1 = \int_{h-\tau}^h ds = \tau$

and the matrix becomes

$$\overline{\phi} = \begin{bmatrix} \phi - \Gamma_0(\tau)L & \Gamma_1(\tau) \\ -L & 0 \end{bmatrix} = \begin{bmatrix} 1 - hL + \tau & \tau \\ -L & 0 \end{bmatrix}$$

Example

We are now interested in computing the eigenvalues of $\overline{\phi}$:

$$\det\left(\overline{\phi} - \lambda I\right) = 0 \Leftrightarrow (1 - hL + \tau L - \lambda)(-\lambda) + \tau L = 0 \Leftrightarrow$$

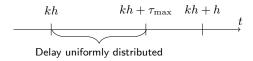
$$\Leftrightarrow \lambda^2 - \lambda (1 - hL + \tau L) + \tau L = 0 \Leftrightarrow$$

$$\Leftrightarrow \lambda_{1,2} = \frac{1 - hL + \tau L \pm \sqrt{(1 - hL + \tau L)^2 - 4\tau L}}{2}$$

The sampled system state is asymptotically stable iff $|\lambda_1|, |\lambda_2| < 1$.

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- \bullet We now assume that the delay is not constant, but a random variable of maximum value $\tau_{\rm max}$
- ullet Using $au_{
 m max}$ as in the previous case, would give conservative results
- Exploit the randomness to get better results
- Jitter: random delay

When is the system stable despite random delay?

Assuming h=1

$$x(k+1) = \phi x(k) + \Gamma u(k)$$

$$y\left(k\right) = Cx\left(k\right) + Du\left(k\right)$$

Taking the z-transform of this system of equations,

$$zX(z) - X(0) = \phi X(z) + \Gamma U(z)$$

$$Y(z) = CX(z) + DU(z)$$

where $X\left(z\right)=Z\left\{ x\left(k\right)\right\} ,$ $U\left(z\right)=Z\left\{ u\left(k\right)\right\}$ and $Y\left(z\right)=Z\left\{ y\left(k\right)\right\}$

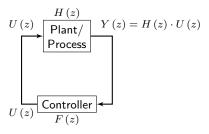
Combining the equations and solving for Y,

$$Y\left(z\right) = C(zI - \phi)^{-1}zX\left(0\right) + \left[C(zI - \phi)^{-1}\Gamma + D\right]U\left(z\right)$$

Definition

Pulse transfer function

$$H(z) \stackrel{\Delta}{=} \left[C(zI - \phi)^{-1} \Gamma + D \right]$$



Assume that the control law is given as

$$u\left(t\right)=f\left(t\right)\ast y\left(t\right)\xrightarrow{Z\left\{ \right\} }U\left(z\right)=F\left(z\right)\cdot Y\left(z\right)$$

 $F\left(z\right)$ can be properly designed in order to affect the controller's decision

Theorem

Consider the WSNCS with a random uniform distributed delay with $\tau_{\max} \leq h$ and $U(z) = F(z) \cdot Y(z)$. The system is stable if

$$\left| \frac{F(z) H(z)}{1 + F(z) H(z)} \right| \le \frac{1}{\tau_{\text{max}} \cdot |z - 1|} \quad \forall \omega$$

where $z = e^{i\omega}$

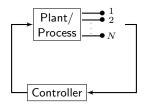
- Sufficient but not necessary condition for stability
- Modify $F\left(z\right)$ and/or au_{max} to achieve stability

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WSNCS with rate constraint

Consider a WSNCS where the same state, $x\left(k\right)$, is observed at different time instants with different measurement functions



The system is described by the set of difference equations

$$x(k+1) = f_i(x(k))$$
 $i = 1, ..., N$

where $1, \ldots, N$ the set of discrete states with the respective associated rates r_1, r_2, \ldots, r_N . These rates are the fraction of time that each discrete state occurs, that is

$$r_i = \frac{1}{t_i} \quad i = 1, ..., N$$

WSNCS with rate constraint

The stability condition of such model is given by the following theorem

Theorem

Given a WSNCS as defined above, if there exists a Lyapunov function such that $V(x(k)): \mathbb{R}^n \to \mathbb{R}_+$ and scalars $\alpha_1, \alpha_2, ..., \alpha_N$ corresponding to each rate such that

$$\alpha_1^{r_1}\cdot\alpha_2^{r_2}\cdot\dots\cdot\alpha_N^{r_N}>\alpha>1$$

and

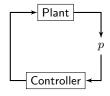
$$V(x(k+1)) - V(x(k)) \le (\alpha_i^{-2} - 1)V(x(k))$$
 $i = 1, ..., N$

then the WSNCS remains exponentially stable with decay rate greater than α

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Packet losses



The input u(t) will depend on whether a packet drop has occurred or not and is given by: $u(t)=\overline{x}(kh)$ for $t\in [kh,kh+h)$, where

$$\overline{x}(kh) = \begin{cases} x(kh-h) & \text{the k:th packet was lost} \\ x(kh) & \text{otherwise} \end{cases}$$

The probability of a packet loss is given by the **packet loss probability** p

WSNCS with packet losses

The characteristic equations of the closed loop system are

$$x((k+1)h) = \phi x(kh) + \Gamma u(kh)$$
$$u(kh) = -L\bar{x}(kh)$$

where recall that

$$\bar{x}\left(kh\right) = \left\{ \begin{array}{ll} x\left(kh\right) & \text{if no packet losses} \\ \bar{x}\left(\left(k-1\right)h\right) & \text{otherwise} \end{array} \right.$$

When is the system stable?

WSNCS with packet losses

Theorem

Suppose that the closed loop system is stable (i.e., $\rho(\phi-\Gamma L)<1$, namely the matrix is stable) when there are not packet losses. Then

- 1. If the open loop system is stable (i.e., $\rho(\phi)$ is stable) then the closed loop system is stable for every p
- 2. If ϕ is unstable, then the closed loop system (with packet losses) is stable when

$$\frac{1}{1 - \frac{\gamma_1}{\gamma_2}} < 1 - p$$

where $\gamma_1 = \log \lambda_{\max}^2 (\phi - \Gamma L)$, $\gamma_2 = \log \lambda_{\max}^2 (\phi)$ and λ_{\max} the maximum eigenvalue of the corresponding matrix

We have the following design parameters:

- ullet The packet loss probability p, that can be changed by the PHY, MAC, routing
- ullet The controller we are assuming L

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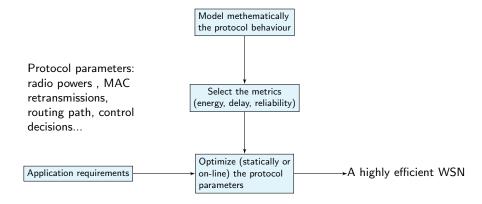
WSNCS design

Energy consumption $E\left(x\right)$

$$\begin{aligned} & \min_{x} E\left(x\right) \\ & \text{s.t. } \Pr\left(\text{succ}\right) \geq 1 - p \\ & \Pr\left(\text{delay} \leq \tau_{\max}\right) \geq \delta \end{aligned}$$

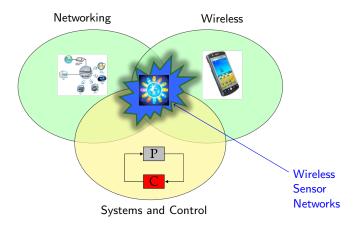
 \boldsymbol{x} collects the protocol and control parameters

WSNs design



The role of mathematical modeling and optimization is central

Wireless Sensor Networks



Summary

- We saw that there is no need to design WSNs that minimize the delay and maximize the packet reception probability
- The controllers can tolerate a certain degree of delay and packet losses
- The efficient design of a wireless sensor network control system can be posed by optimization problems

Exam, October 24-th, 14:00-19:00

- 5 exercises chosen on every part of the course, inspired from the exercises of the compendium and homework
- 5 hours to complete the exam
- Allowed to bring PRINTED lecture slides and draft book, and basic books on math, e.g., Mathematics Handbook by Råde & Westergren
- Not allowed to bring exercise lecture notes
- Not allowed to bring compendium with exercises and solutions
- Results available after 1-2 weeks

Master thesis projects

- Theoretical, practical, or business oriented
- Conduct forefront research
- Possible collaboration with industry
- Interaction with Professors, Research Associates, and PhD students
- You can propose the topic, or ask for a project on
 - Distributed optimization over WSNs
 - Distributed detection and estimation
 - Design of wireless sensor networked control systems
 - Future wireless networks
 - Internet of Things
 - MAC, Routing
 - Smart grids
 - Privacy

PhD in Electrical Engineering

- For motivated and hard working students with high grades (i.e., talented students), possibility of pursuing a PhD
- Pretty high salary for studying
- International collaborations and travel, UC Berkeley, Stanford University, MIT, Caltech,...
- Competitive
- World-wide job market
- Research (50%), courses (30%), teaching (20%) = fun (100%)
- 4-5 years to earn the PhD

Some success stories...









- Pangun Park, PhD 2011 on WSNs, took my master thesis project
 - Admitted to the PhD program at KTH EE School in 2007
 - Research Associate at University of California at Berkeley, Electrical Engineering and Computer Sciences Department, Thrust Center (2011-2013)
- Piergiuseppe Di Marco, PhD on WSNs in 2013, took my master thesis project
 - In 2012 was for 6 months at University of California at Berkeley, Electrical Engineering and Computer Sciences Department, DOP Center
 - Now Experienced Researcher of IoT in Ericsson Research, Stockholm

This afternoon

- Project presentations: 5 minutes (sharp!) per group + 2 minutes questions
 - ▶ What is the topic
 - Why the topic is important in relation to the overall WSN
 - What are the key aspects of the topic you studied
 - What has been implemented and experimental results