

# Principles of Wireless Sensor Networks

<https://www.kth.se/social/course/EL2745/>

## Lecture 12

# Wireless Sensor Network Control Systems 1

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# Course content

- Part 1

- ▶ Lec 1: Introduction to WSNs
- ▶ Lec 2: Introduction to Programming WSNs

- Part 2

- ▶ Lec 3: Wireless Channel
- ▶ Lec 4: Physical Layer
- ▶ Lec 5: Medium Access Control Layer
- ▶ Lec 6: Routing

- Part 3

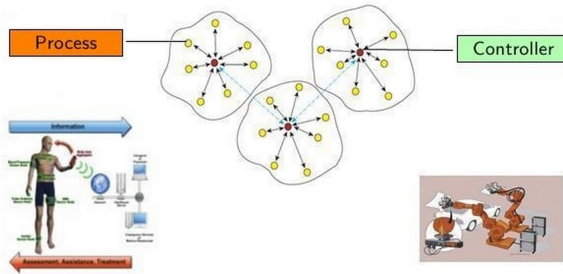
- ▶ Lec 7: Distributed Detection
- ▶ Lec 8: Static Distributed Estimation
- ▶ Lec 9: Dynamic Distributed Estimation
- ▶ Lec 10: Positioning and Localization
- ▶ Lec 11: Time Synchronization

- Part 4

- ▶ Lec 12: Wireless Sensor Network Control Systems 1
- ▶ Lec 13: Wireless Sensor Network Control Systems 2

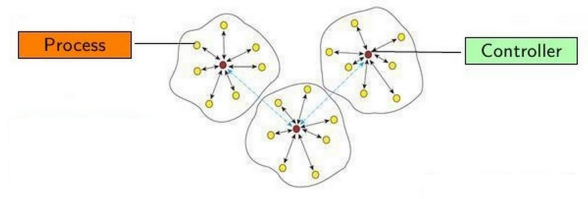
# Previous lecture

Application
Presentation
Session
Transport
Routing
MAC
Phy



How to synchronize nodes?

# Today's learning goals



- How the process state dynamics over time are mathematically modeled?
- How such state dynamics can be controlled by closing the loop  
process→controller→process?
- How to discretize the continuous time model of the dynamics?
- What is the concept of state stability of closed loop control systems?

# Outline

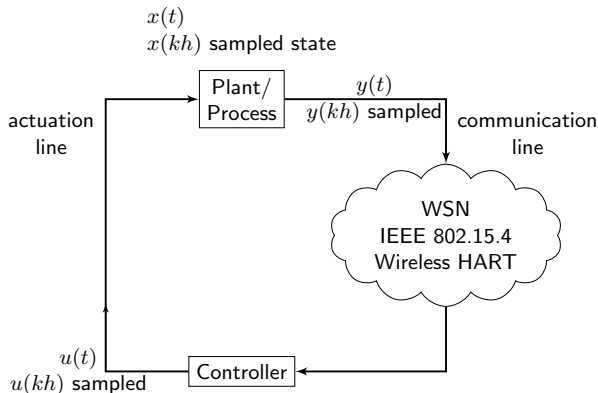
- Wireless Sensor Network Control Systems (WSNCS)
- State space description of a control system
- Stability and asymptotic stability of a control system

# Outline

- Wireless Sensor Network Control Systems (WSNCS)
- State space description of a control system
  - ▶ Linear model
    - Continuous time description
    - Discretization of state space model
  - ▶ Non-linear model
- Stability and asymptotic stability of a control system

# Wireless Sensor Network Control Systems (WSNCS)

Closed-loop system



# Wireless Sensor Network Control Systems

$k$ : discrete time

$h$ : sampling interval

- $u(kh)$ : control decision
- $x(kh)$ : state of the process/plant
- $y(kh)$ : output of the state (measured by sensors)
  
- The **GOAL** of the controller is to bring the state  $x(kh)$  in a desired region by taking measurements  $y(kh)$  and a control decision  $u(kh)$
- Delay and packet loss probability affect the way the measurements  $y(t)$  are received in the controller

This lecture gives the basic control theory background for WSNCS. The effect of the network on the controller is studied next lecture

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# Continuous time description

Let  $x(t)$  be the state of the system (temperature, position, ...)

We assume that the physical process is described by the time-invariant state space model

## Linear model

$$\frac{dx(t)}{dt} \triangleq \dot{x}(t) = Ax(t) + Bu(t) \quad \text{state model} \quad (1)$$

$$y(t) = Cx(t) + Du(t) \quad \text{measurement model}$$

where  $A, B, C, D$  are assumed to be known matrices

# Continuous time description

Assuming that  $x(kh)$  is known, then the solution of the differential equation (1) is

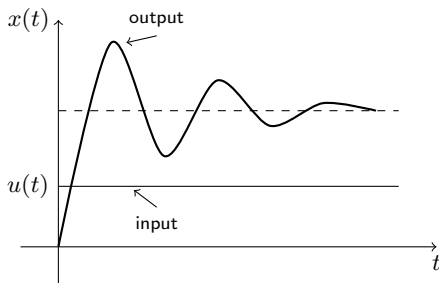
$$x(t) = e^{A(t-kh)} \cdot x(kh) + \int_{kh}^t e^{A(t-\tau)} B u(\tau) d\tau \quad t > kh \quad (2)$$

The control decision  $u(t)$  can be properly chosen to bring the system state  $x(t)$  in a desired region.

## Example: "The step response"

Suppose  $x(0) = 0$  and  $x(t) \in \mathbb{R}$

The step response is defined as the evolution of the state  $x(t)$ , i.e., the solution of (1), for an input  $u(t) = \begin{cases} 0 & t \leq 0 \\ 1 & t > 0 \end{cases}$



The state may evolve to a stabilized condition after possible oscillations

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# Discretization of state space model

Assume  $u(t)$  constant in the interval  $kh \leq t \leq kh + h$

Then, (2) becomes

$$\begin{aligned}x(t) &= e^{A(t-kh)} \cdot x(kh) + \int_{kh}^t e^{A(t-\tau)} d\tau B u(t) = \\&= e^{A(t-kh)} \cdot x(kh) + \int_0^{t-kh} e^{A\tau} d\tau B u(kh) = \phi_t x(kh) + \Gamma_t u(kh)\end{aligned}$$

Let  $t = kh + h$

$$x(kh + h) = \phi x(kh) + \Gamma u(kh) \quad (3)$$

$$\text{where } \phi = e^{Ah} \text{ and } \Gamma = \int_0^h e^{A\tau} d\tau B$$

# Discretization of state space model

Recursively from (3),

$$x(kh + 2h) = \phi x(kh + h) + \Gamma u(kh + h)$$

Therefore, the solution of (3), given  $x(0)$  and  $u(kh) \forall k$ , is

$$x(kh) = \phi^k x(0) + \sum_{j=0}^{k-1} \phi^{k-1-j} \Gamma u(jh)$$

Note that the matrix exponential  $e^{Ah}$  can be expressed equivalently as a power series

$$e^{Ah} = I + Ah + \frac{A^2 h^2}{2} + \dots$$

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# Non-linear model of the state

## Observation

Control decision is chosen as a function of the state

$$u(t) = f(x(t))$$

Therefore, consider a state that evolves according to a non-linear law

$$\dot{x}(t) = a(x(t))$$

$$y(t) = c(x(t))$$

where  $a$  and  $c$  are be non-linear functions in general

What is the solution of that system?

# Non-linear model of the state

## Non-linear differential equation

$$x(t + kh) = x(t) + \int_t^{t+kh} a(x(\tau)) d\tau$$

- In general, the integral difficult to solve.
- However, by deriving opportune upper bounds of  $a(x(t))$ , it is possible to show important properties of the system (e.g., stability).

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# Stability

Let us consider the discrete-time differential equation

$$x(kh + h) = g(x(kh)) \quad (4)$$

where  $g$  can be linear or non-linear

## Definition

A specific solution of (4),  $x^*(kh)$ , is called stable,  
if  $\forall \varepsilon > 0 \quad \exists \delta(\varepsilon) : \forall$  other solution  $x(kh)$

$$\|x(0) - x^*(0)\| \leq \delta \Rightarrow \|x(kh) - x^*(kh)\| \leq \varepsilon \quad \forall k$$

# Asymptotic stability

We consider the same equation as on the previous slide:

$$x(kh + h) = g(x(kh)) \quad (5)$$

where  $g$  can be linear or non-linear.

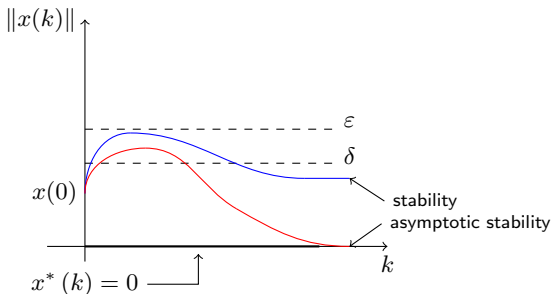
## Definition

A specific solution  $x^*(k)$  of (5) is called **asymptotically stable** if it is stable and if there is a  $\delta > 0$  such that for every other solution  $x(k)$  it holds that:

$$\|x(0) - x^*(0)\| \leq \delta \Rightarrow \|x(k) - x^*(k)\| \rightarrow 0 \quad \text{as } k \rightarrow \infty$$

# Example 1

Assume that  $x^*(kh) = 0$  is a solution of (5). The figure shows the typical behaviour of other solutions in case  $x^*(kh)$  is stable or asymptotically stable.



# The linear case

Consider a linear difference equation

$$x(kh + h) = A \cdot x(kh) \quad (6)$$

where  $A$  known matrix  $\in \mathbb{R}^{n \times n}$

## Definition

A linear difference equation of the form (6) is (asymptotically) stable if the constant solution  $x^*(k) = 0$  is (asymptotically) stable.

How do we choose matrix  $A$  in order to have

1. stability?
2. asymptotic stability?

# The linear case

The answer is given by the following theorem:

## Theorem (Stability of linear difference equations)

Let  $\rho(A) = \max\{|\lambda|, \lambda \text{ is an eigenvalue of } A\}$ .

- (i)  $x(kh + h) = A \cdot x(kh)$  is stable if and only  $\rho(A) \leq 1$ .
- (ii)  $x(kh + h) = A \cdot x(kh)$  is asymptotically stable if and only  $\rho(A) < 1$ .

# The linear case: intuition

1.  $\|x(0) - 0\| \leq \delta \Rightarrow \|x(kh)\| \leq \varepsilon \quad ?$

$$\|x(kh)\| = \|A \cdot x((k-1)h)\| = \dots = \|A^k \cdot x(0)\| \leq \|A^k\| \cdot \|x(0)\| \leq \|A\|^k \cdot \delta$$

To achieve stability, choose matrix  $A$  that does not grow with  $k$

This is when the maximum absolute eigenvalue of  $A$ ,  $\rho(A) \leq 1$

2.  $\lim_{k \rightarrow \infty} \|x(kh)\| \leq \lim_{k \rightarrow \infty} \|A\|^k \cdot \delta \rightarrow 0$

In this case,  $\rho(A) < 1$

- In scalar case,  $A$  is constant and  $\rho(A) = A$
- If the eigenvalues are larger than 1  $\Rightarrow$  instability

# Summary

- We have seen the basic aspects of control systems
  - ▶ Mathematical description of the state evolution
  - ▶ Discretization
  - ▶ Stability

# Next lecture

- WSNCS, robustness to packet delays and losses