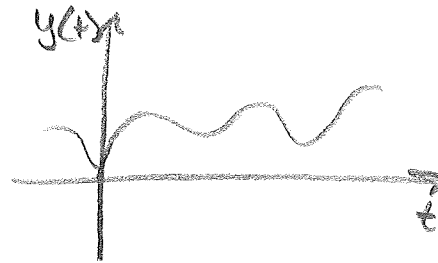


Theory:

Norms: Good for comparing signals & systems.

Signal norms:

$y(t)$ - a signal



How "big" is this signal?

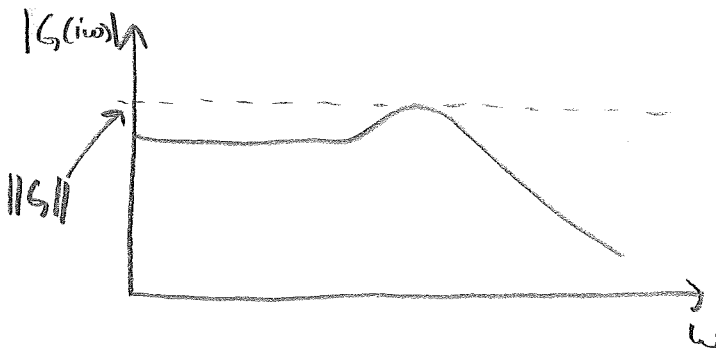
$$\|y\|_2 = \left(\int_{-\infty}^{\infty} |y(t)|^2 dt \right)^{1/2} \quad (\text{energy})$$

$$\|y\|_{\infty} = \sup_t |y(t)| \quad (\text{peak value})$$

System norm: How much "bigger" can the system make a signal

Linear systems: H_{∞} -norm

$$\|G\| := \sup_{\omega} |G(i\omega)| = \| |G(i\omega)| \|_{\infty}$$



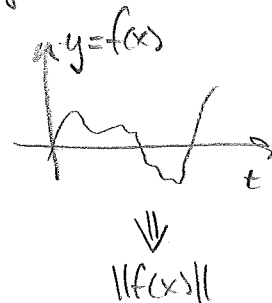
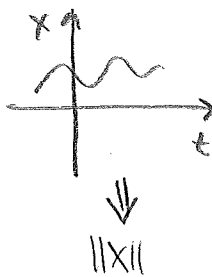
Peak in bode-diagram.

Static nonlinearity:

$$y = f(x)$$

$$\|f\| := \sup_x \frac{\|f(x)\|_2}{\|x\|_2}$$

x & y are signals

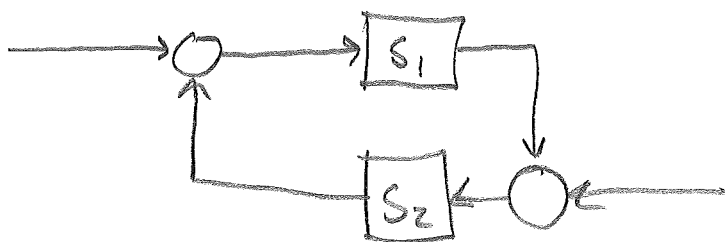


Useful inequalities

$\|A+B\| \leq \|A\| + \|B\|$ (triangle ineq.) Valid for all norms

$\|A \cdot B\| \leq \|A\| \cdot \|B\|$ (sub-multiplicative) Valid for some norms, e.g. H_{∞} .

Small gain theorem:



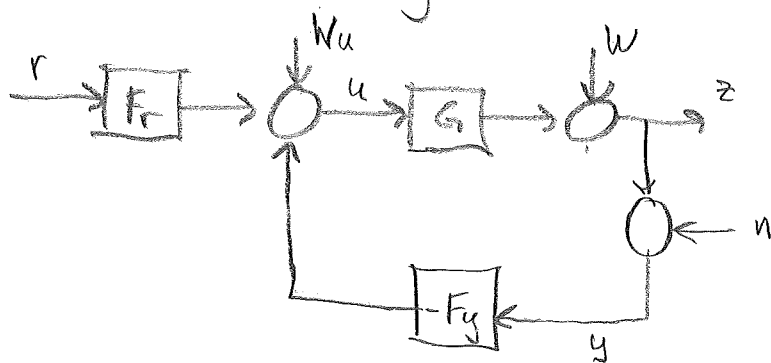
If S_1 & S_2 are stable and

Linear case
 $\|S_1, S_2\| < 1$

nonlinear case
 $\|S_1\| \cdot \|S_2\| < 1$

then the closed loop is stable.

Internal stability



A system is internally stable if all signals remain bounded after injecting a bounded disturbance anywhere in the system.

⇒ Need to look at "all" transfer functions from $r, W_u, W, n \rightarrow z, y, u$

Theorem: (SISO)

Internally stable if and only if

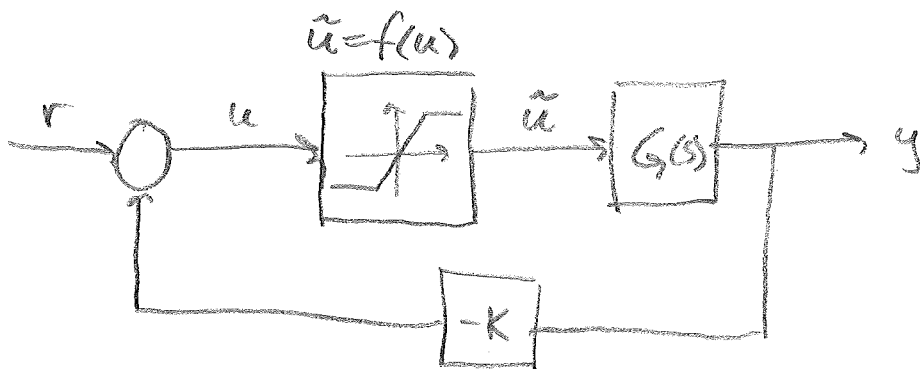
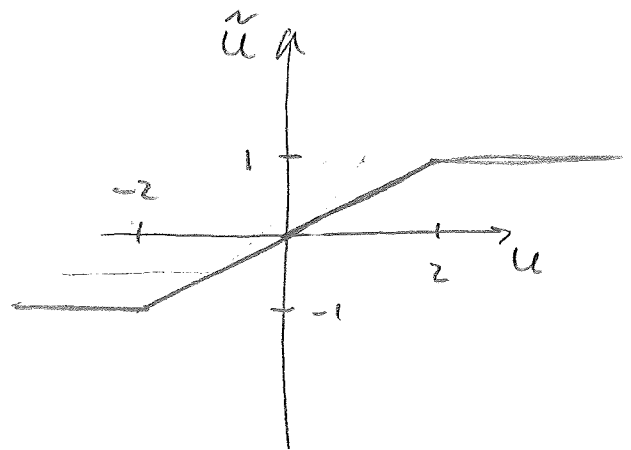
$$S = \frac{1}{1+F_y G}, \quad S G = \frac{G}{1+F_y G}, \quad S F_y = \frac{F_y}{1+F_y G}, \quad F_r$$

are all stable.

1,3

$$G(s) = \frac{2}{s^2 + 2s + 2}$$

$$\tilde{u} = \begin{cases} 1 & u > 2 \\ \frac{1}{2}u & |u| < 2 \\ -1 & u < -2 \end{cases} \Rightarrow$$



For which values of K are the system stable according to the small gain theorem?

Since we have a nonlinearity we need to use the "non-linear version" of the small-gain theorem

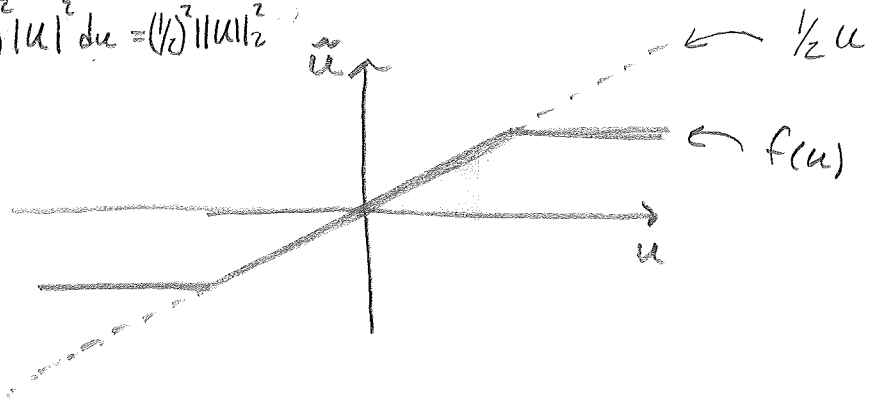
$\|f\| \cdot \|G\| \cdot \|K\| < 1 \Rightarrow$ stable. if all subsystems stable.

Is f input-output stable? Yes, since the output is bounded $|f(u)| < 1 \forall u$.

$\|f\|$: We note that $|\hat{u}| = |f(u)| \leq \frac{1}{2}|u| \Rightarrow$

$$\|\hat{u}\|_2^2 = \int_{-\infty}^{\infty} |\hat{u}|^2 du \leq \int_{-\infty}^{\infty} \left(\frac{1}{2}\right)^2 |u|^2 du = \left(\frac{1}{2}\right)^2 \|u\|_2^2$$

$$\Rightarrow \|f\| \leq \frac{1}{2}$$



$\|G\|$: From the definition we have

$$\|G\| = \sup_{\omega} |G(i\omega)|$$

$$|G(i\omega)| = \frac{z}{|i^2\omega^2 + 2i\omega + z|} = \frac{z}{\sqrt{4\omega^2 + (z - \omega^2)^2}}$$

Maximizing $|G(i\omega)|$ is the same as minimizing $4\omega^2 + (z - \omega^2)^2$

$$\frac{d}{d\omega} 4\omega^2 + (z - \omega^2)^2 = 8\omega - 4\omega(z - \omega^2) = 4\omega^3$$

We see that this is only zero at $\omega = 0 \Rightarrow$

$$\|G\| = |G(i0)| = 1$$

Alternative: If we rewrite the system as

$$G(s) = \frac{\omega_0^2}{s^2 + 2\zeta\omega_0 s + \omega_0^2} \Rightarrow \zeta = \frac{1}{\sqrt{2}} \Rightarrow$$

system is well damped second order system.

$$\Rightarrow \|G\| = \sup_{\omega} |G(i\omega)| = G(0) = 1$$

$\|K\|$:

The gain of a constant is the constant

$$\|K\| = |K|$$

For stability we then get by the S.G.T.

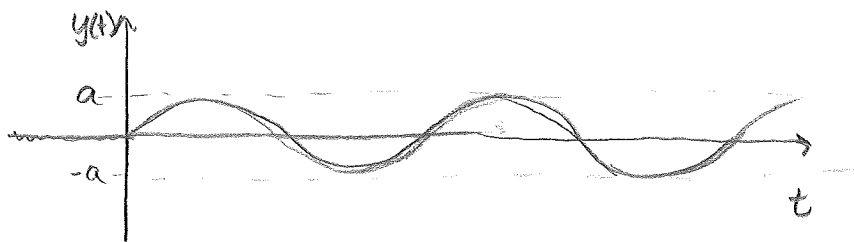
$$\|f\| \cdot \|G\| \cdot \|K\| = \frac{1}{2} \cdot 1 \cdot |K| < 1 \Rightarrow \underline{\underline{|K| < 2}}$$

Exercise 1,4

Calculate $\|\cdot\|_\infty$ & $\|\cdot\|_2$ for the signals

$$a) \quad y(t) = \begin{cases} a \sin(t) & t > 0 \\ 0 & t \leq 0 \end{cases}$$

$$\|y(t)\|_\infty = \sup_t |y(t)| = \sup_t |a \sin(t)| = |a| \underbrace{\sup_t |\sin(t)|}_1 = |a|$$



$$\|y(t)\|_2 = \sqrt{\int_{-\infty}^{\infty} y(t)^2 dt} = \sqrt{\int_0^{\infty} a^2 \sin^2 t dt} =$$

$$= |a| \sqrt{\int_0^{\infty} \sin^2 t dt} = |a| \sqrt{\int_0^{\infty} \frac{1 - \cos 2t}{2} dt}$$

$$= |a| \sqrt{\int_0^{\infty} \frac{1}{2} dt - \underbrace{\int_0^{\infty} \frac{1}{2} \cos 2t dt}_{< 1 \forall t > 0}} > |a| \sqrt{\infty - 1} = \infty$$

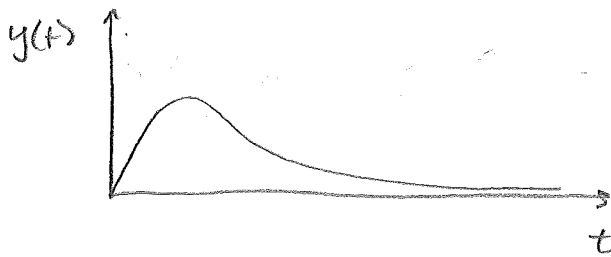
$$b) \quad y(t) = \begin{cases} 1/t & t > 1 \\ 0 & t \leq 1 \end{cases}$$



$$\|y(t)\|_\infty = \sup_{t>1} |1/t| = 1$$

$$\|y(t)\|_2 = \sqrt{\int_1^{\infty} 1/t^2 dt} = \sqrt{\left[\frac{-1}{t} \right]_1^{\infty}} = \sqrt{0 - (-1)} = 1$$

$$c_1 \quad y(t) = \begin{cases} e^{-t}(1-e^{-t}) & t > 0 \\ 0 & t \leq 0 \end{cases}$$



$$\|y(t)\|_{\infty} = \sup_{t > 0} e^{-t}(1-e^{-t}) = \sup_{t > 0} e^{-t} - e^{-2t}$$

We take the derivative and equate to zero in order to find the max.

$$\frac{dy}{dt} = -e^{-t} + 2e^{-2t} = 0 \quad \Leftrightarrow \quad \underbrace{e^{-t}}_{> 0 \forall t} (2e^{-t} - 1) = 0$$

$$2e^{-t} = 1 \quad \Leftrightarrow \quad \log(2e^{-t}) = 0$$

$$\Leftrightarrow \log 2 + \log(e^{-t}) = 0$$

$$\Leftrightarrow t = \log 2$$

So we get the maximum at $t = \log(2)$:

$$\begin{aligned} y(\log(1/2)) &= e^{-\log(2)} - e^{-2\log(2)} = e^{\log(1/2)} - e^{\log(1/4)} = \\ &= 1/2 - 1/4 = 1/4 \end{aligned}$$

$$\begin{aligned} \|y(t)\|_2 &= \sqrt{\int_0^{\infty} (e^{-t} - e^{-2t})^2 dt} = \\ &= \sqrt{\int_0^{\infty} (e^{-2t} - 2e^{-3t} + e^{-4t}) dt} = \sqrt{\left[-\frac{1}{2}e^{-2t} + \frac{2}{3}e^{-3t} - \frac{1}{4}e^{-4t} \right]_0^{\infty}} = \\ &= \sqrt{\frac{1}{2} - \frac{2}{3} + \frac{1}{4}} = \frac{1}{\sqrt{12}} \end{aligned}$$

1.5

$$\|G\| = \| |G(i\omega)| \|_{\infty} = \sup_{\omega} |G(i\omega)|$$

$$|G(i\omega)| = \left| \frac{\omega_0^2 \overset{\text{constant}}{}}{i^2 \omega^2 + 2\zeta \omega_0 i \omega + \omega_0^2} \right| = \frac{\omega_0^2}{\sqrt{(\omega_0^2 - \omega^2)^2 + 4\zeta^2 \omega_0^2 \omega^2}}$$

Maximizing $|G(i\omega)|$ is the same as
minimizing $(\omega_0^2 - \omega^2)^2 + 4\zeta^2 \omega_0^2 \omega^2$

$$\begin{aligned} \frac{d}{d\omega} (\omega_0^2 - \omega^2)^2 + 4\zeta^2 \omega_0^2 \omega^2 &= 2(\omega_0^2 - \omega^2) \cdot -2\omega + 8\zeta^2 \omega_0^2 \omega = \\ &= 4\omega (2\zeta^2 \omega_0^2 - \omega_0^2 + \omega^2) = \\ &= 4\omega (\omega_0^2 (2\zeta^2 - 1) + \omega^2) = 0 \end{aligned}$$

We get the solutions

$$\omega = 0$$

$$\omega = (\pm) \omega_0 \sqrt{1 - 2\zeta^2}$$

We require a real positive solution so
we need $\zeta < \frac{1}{\sqrt{2}}$

We get two cases:

(I) $0 < \zeta < \frac{1}{\sqrt{2}}$, a partly damped system

$$\begin{aligned} \|G\| &= |G(i\omega_0 \sqrt{1 - 2\zeta^2})| = \frac{\omega_0^2}{\sqrt{(\omega_0^2 - \omega_0^2(1 - 2\zeta^2))^2 + 4\zeta^2 \omega_0^4 (1 - 2\zeta^2)}} = \\ &= \frac{\omega_0^2}{\sqrt{4\omega_0^4 \zeta^2 (1 - 2\zeta^2)}} = \frac{1}{2\zeta \sqrt{1 - 2\zeta^2}} > 1 \end{aligned}$$

(II) $\zeta > \frac{1}{\sqrt{2}} \Rightarrow \|G\| = |G(i0)| = \frac{\omega_0^2}{0 + 0 + \omega_0^2} = 1$

1.6

Small gain theorem:

Both systems need to be stable
so we must have $a > 0$.

By the the S.G.T. we need

$$\|K \cdot \frac{a}{s+a}\| < 1 \text{ for stability} \Rightarrow$$

$$\sup_w \frac{Ka}{\sqrt{w^2+a^2}} < 1$$

To maximize this we want

w^2+a^2 as small as possible \Rightarrow

$$w=0$$

$$\text{We get } \|K \cdot \frac{a}{s+a}\| = \left| \frac{Ka}{a} \right| = |K|$$

Hence $|K| < 1$ & $a > 0$ for stability

Poles of Closed-loop system:

The closed loop-system becomes

$$G_c(s) = \frac{K \frac{a}{s+a}}{1 - K \frac{a}{s+a}} = \frac{Ka}{s+a(1-K)}$$

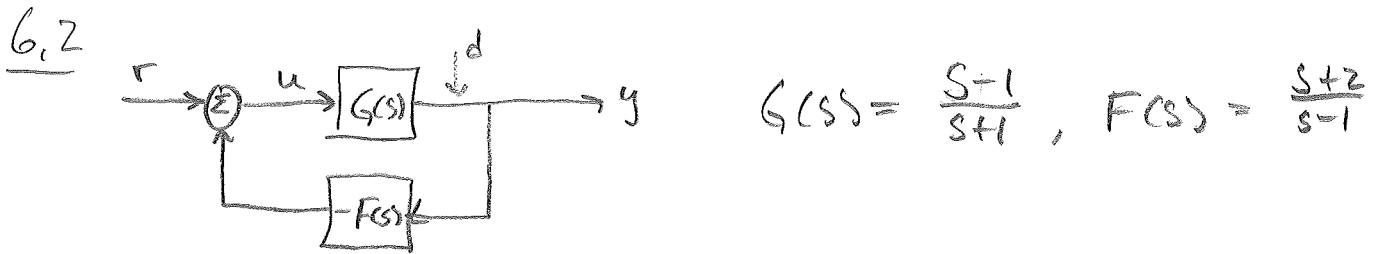
attractive
poles

Case 1: $a > 0$ We see that we need

$K < 1$ for stability (negative pole)

Case 2: $a < 0$

We need $K > 1$ for stability.



Compute S , T & G_c

$G_c(s): Y(s) = G_c(s) R(s)$

$$Y(s) = G(s) R(s) - G(s) F(s) Y(s)$$

$$\Rightarrow G_c(s) = \frac{G(s)}{1 + G(s)F(s)} = \frac{\frac{s-1}{s+1}}{1 + \frac{s+2}{s-1} \frac{s-1}{s+1}} = \frac{s-1}{2s+3}$$

$S: L(s) = F(s)G(s)$

$$S(s) = \frac{1}{1+L(s)} = \frac{1}{1 + \frac{s+2}{s-1} \frac{s-1}{s+1}} = \frac{s+1}{2s+3}$$

$T: T = 1 - S = \frac{2s+3}{2s+3} - \frac{s+1}{2s+3} = \frac{s+2}{2s+3}$

All of these are stable. However, consider the effect of a disturbance on the control signal.

$G_{du}: U(s) = -F(s)G(s)U(s) - F(s)D(s)$

\Rightarrow

$$G_{du}(s) = \frac{-F(s)}{1 + F(s)G(s)} = -F(s)S(s) = \frac{s+2}{s-1} \frac{s+1}{2s+3}$$

Hence, the control is unstable and grows unbounded for small disturbances.

We are not internally stable.