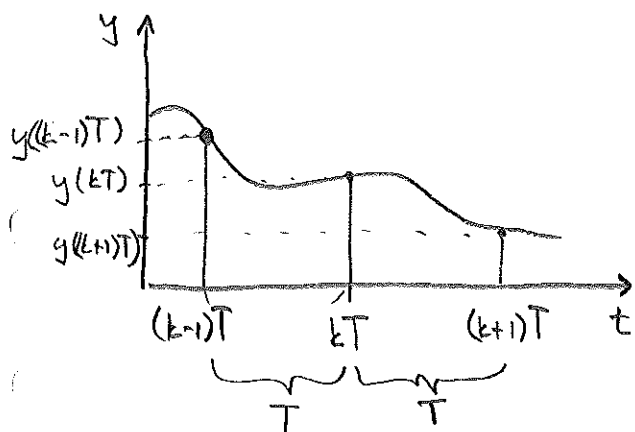


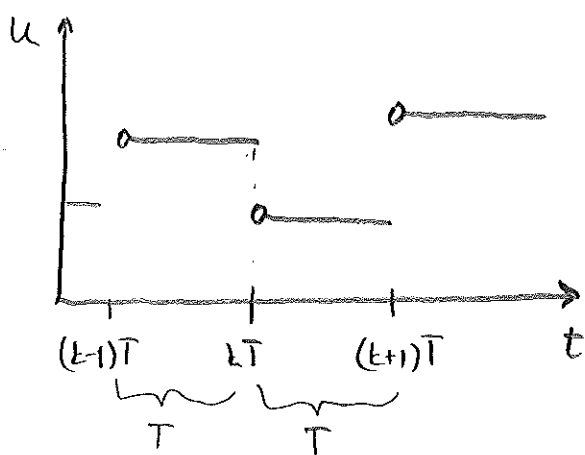
Theory

Discrete time:

Consider the system at discrete points in time.



Assumes u piecewise constant



Notation: $u_k = u(kT)$ $u_{k+1} = u((k+1)T)$
 $y_k = y(kT)$

$x_k = x(kT)$
 $(x_k)_1 = x_1(kT)$ i.e. $(x_k)_1$ $\begin{bmatrix} x_1(kT) \\ x_2(kT) \\ x_3(kT) \end{bmatrix}$
 $(x_k)_3$

Difference equations

$$x_{k+1} = F x_k + G u_k$$

$$y_k = H x_k$$

Sampling: Continuous \rightarrow Discrete

$$\dot{x} = Ax + Bu$$

$$y = Cx$$

Sample with interval $T \Rightarrow$

$$x_{k+1} = F x_k + G u_k$$

$$y_k = C x_k$$

$$F = e^{AT}, \quad G = \int_0^T e^{A(T-t)} B dt, \quad H = C$$

Matrix exponential:

$$e^{At} = I + \sum_{k=1}^{\infty} \frac{A^k t^k}{k!}$$

alternative

$$e^{At} = \mathcal{L}^{-1} \{ (sI - A)^{-1} \}$$

Stability: $\lambda_i = \text{eigenvalue of } F$

stable if $|\lambda_i| < 1 \forall i$

Observable: Same as for cont. sys.

observable if $O = \begin{bmatrix} H \\ HF \\ \vdots \\ HF^{n-1} \end{bmatrix}$ has full rank

Controllable:

Controllable if $S = [G \quad FG \quad F^2G \quad \dots \quad F^{n-1}G]$ has full rank.

Model Predictive Control (MPC):

We can predict future outputs as functions of future inputs.

Ex:

$$\hat{y}_{k+1} = Hx_{k+1} = H(Fx_k + Gu_k) = HFx_k + HGx_k$$

Idea: minimize cost J involving future outputs.

$$1) \text{ solve } \min_u J(u, \hat{y}) \quad u = \begin{pmatrix} u_k \\ u_{k+1} \\ \vdots \\ u_{k+N} \end{pmatrix}$$

$$\text{S.t. } \begin{aligned} x_{k+1} &= Fx_k + Gu_k \\ y_k &= Hx_k \end{aligned} \quad \hat{y} = \begin{pmatrix} \hat{y}_{k+1} \\ \vdots \\ y_{k+N} \end{pmatrix}$$

\Rightarrow optimal control sequence u_k

2) apply first control u_k and go to 1)

Note: Can include hard constraints such as
 $|u| < 1$ or $|y| < 1$

MPL is computationally expensive.

4,1

The cont. system

$$\dot{x}(t) = \underbrace{\begin{pmatrix} 0 & \omega \\ -\omega & 0 \end{pmatrix}}_A x(t) + \underbrace{\begin{pmatrix} 1 \\ 0 \end{pmatrix}}_B u(t)$$

$$y(t) = \underbrace{(1 \ 0)}_C x(t)$$

is sampled with interval T . Determine the corresponding discrete time system. Is the sampled system observable?

i) find F , G & H .

$$F = e^{AT} = \mathcal{L}^{-1}\{(sI - A)^{-1}\} = \mathcal{L}^{-1}\left\{\begin{pmatrix} s & \omega \\ -\omega & s \end{pmatrix}^{-1}\right\} =$$

$$= \mathcal{L}^{-1}\left\{\frac{1}{s^2 + \omega^2} \begin{pmatrix} s & -\omega \\ \omega & s \end{pmatrix}\right\} = \begin{pmatrix} \cos(\omega T) & \sin(\omega T) \\ -\sin(\omega T) & \cos(\omega T) \end{pmatrix}$$

$$G = \int_0^T e^{At} B dt = \int_0^T \begin{pmatrix} \cos(\omega t) & \sin(\omega t) \\ -\sin(\omega t) & \cos(\omega t) \end{pmatrix} \begin{pmatrix} 1 \\ 0 \end{pmatrix} dt =$$

$$= \int_0^T \begin{pmatrix} \cos(\omega t) \\ -\sin(\omega t) \end{pmatrix} dt = \left[\frac{1}{\omega} \begin{pmatrix} \sin(\omega t) \\ \cos(\omega t) \end{pmatrix} \right]_0^T = \frac{1}{\omega} \begin{pmatrix} \sin(\omega T) \\ \cos(\omega T) - 1 \end{pmatrix}$$

$$H = C$$

The system is thus

$$x_{k+1} = \begin{pmatrix} \cos(\omega T) & \sin(\omega T) \\ -\sin(\omega T) & \cos(\omega T) \end{pmatrix} x_k + \frac{1}{\omega} \begin{pmatrix} \sin(\omega T) \\ \cos(\omega T) - 1 \end{pmatrix} u_k$$

$$y_k = (1 \ 0) x_k$$

ii) observable?

The observability matrix is

$$O = \begin{bmatrix} H \\ H F \end{bmatrix} = \begin{pmatrix} 1 & 0 \\ \cos(\omega T) & \sin(\omega T) \end{pmatrix}$$

Full rank if $\sin(\omega T) \neq 0 \Rightarrow$

$$\omega T \neq n \cdot \pi \quad n \in \mathbb{Z}$$

$$T \neq \frac{n\pi}{\omega} \quad n \in \mathbb{Z} \Rightarrow \text{observable.}$$

16.2 Given the discrete time system

$$\begin{aligned}
 x_{k+1} &= Fx_k + Gu_k & F &= \begin{pmatrix} 1.7 & 1 \\ -0.7 & 0 \end{pmatrix} & G &= \begin{pmatrix} 0.9 \\ -0.6 \end{pmatrix} \\
 y_k &= Hx_k & H &= (1 \ 0)
 \end{aligned}$$

Find a control law which solves

$$\min_u J = \min_u \sum_{i=1}^3 \underbrace{(r_{k+i} - \hat{y}_{k+i})^2}_{\text{predicted control error}} + 0.1 \underbrace{(u_{k+i-1})^2}_{\text{control effort}}$$

where \hat{y}_{k+i} is the predicted output i steps ahead.

Rewrite J with vectors:

$$\text{Let } \vec{r} = \begin{pmatrix} r_{k+1} \\ r_{k+2} \\ r_{k+3} \end{pmatrix} \quad \vec{\hat{y}} = \begin{pmatrix} \hat{y}_{k+1} \\ \hat{y}_{k+2} \\ \hat{y}_{k+3} \end{pmatrix} \quad \vec{u} = \begin{pmatrix} u_k \\ u_{k+1} \\ u_{k+2} \end{pmatrix}$$

$$\text{then } J = (\vec{r} - \vec{\hat{y}})^T (\vec{r} - \vec{\hat{y}}) + 0.1 \vec{u}^T \vec{u}$$

Express J in terms of \vec{u} :

$$\hat{y}_{k+1} = Hx_{k+1} = H(Fx_k + Gu_k) = HFx_k + HG u_k$$

$$\begin{aligned}
 \hat{y}_{k+2} &= Hx_{k+2} = H(Fx_{k+1} + Gu_{k+1}) = H(F(Fx_k + Gu_k) + Gu_{k+1}) = \\
 &= HF^2 x_k + HFG u_k + HG u_{k+1}
 \end{aligned}$$

$$\hat{y}_{k+3} = \dots = HF^3 x_k + HF^2 G u_k + HFG u_{k+1} + HG u_{k+2}$$

$$\Rightarrow \vec{y} = \underbrace{\begin{pmatrix} HF \\ HF^2 \\ HF^3 \end{pmatrix}}_P x_k + \underbrace{\begin{bmatrix} HG & 0 & 0 \\ HFG & HG & 0 \\ HF^2G & HFG & HG \end{bmatrix}}_S \vec{u}$$

$$\vec{y} = Px_k + S\vec{u} \quad \Rightarrow$$

$$J = (\vec{r} - Px_k - S\vec{u})^T (\vec{r} - Px_k - S\vec{u}) + 0,1 \vec{u}^T \vec{u}$$

Expand J:

$$J = \vec{r}^T \vec{r} - 2\vec{r}^T Px_k + x_k^T P^T Px_k - 2\vec{r}^T S\vec{u} + 2x_k^T P^T S\vec{u} + \vec{u}^T S^T S\vec{u} + 0,1 \vec{u}^T \vec{u}$$

J Quadratic in $\vec{u} \Rightarrow$ easy to minimize.

$$\frac{\partial J}{\partial \vec{u}} = -2S^T \vec{r} + 2S^T Px_k + 2S^T S\vec{u} + 0,2I\vec{u} = 0$$

$$\Rightarrow \text{the minimizing } \vec{u} = (S^T S + 0,1I)^{-1} S^T (\vec{r} - Px_k)$$

Inserting values for F, G, & H \Rightarrow

$$P = \begin{pmatrix} 1,7 & 1 \\ 2,2 & 1,7 \\ 2,5 & 2,2 \end{pmatrix} \quad S = \begin{pmatrix} 0,9 & 0 & 0 \\ 0,93 & 0,9 & 0 \\ 0,95 & 0,93 & 0,9 \end{pmatrix}$$

We are only interested in u_k (remember that we will recompute u_{k+1}, \dots next iteration)

Doing the multiplications etc. \Rightarrow

$$u_k = -(1.7 \quad 1.1) x_k + (0.9 \quad 0.1 \quad 0.01) \vec{r}$$

We have u_k expressed as a state-feedback.
 We want to express it in terms of the measurements we have access to, (y_k, y_{k-1}, \dots)

Note that

$$(x_k)_1 = y_k \quad (x_k)_2 = (x_{k-1})_1 - 0.6 u_{k-1} = y_{k-1} - 0.6 u_{k-1}$$

current measurement

last measurement

last applied control

The final control law becomes

$$u_k = -1.7 y_k - 1.1 y_{k-1} + 0.64 u_{k-1} + 0.9 r_{k+1} + 0.1 r_{k+2} + 0.01 r_{k+3}$$

Note: need to know future references.

16.4

A system with model

$$y_{k+1} = -y_k + 2u_k$$

is to be controlled by an MPC-controller

$$\min_u \sum_{i=k}^{k+N_p} y_i^2 + \sum_{i=k}^{k+N_p-1} u_i^2$$

$$\text{s.t. } |u| \leq 1$$

Let $N_p=1$ and translate the problem into a quadratic programming problem on the form

$$\min_u u^T H u + h^T u \quad \text{s.t. } L u \leq b$$

i.e. find H , h^T , L & b .

Expand J:

$$J = \sum_{i=k}^{k+1} y_i^2 + \sum_{i=k}^k u_i^2 = y_{k+1}^2 + y_k^2 + u_k^2$$

Use system model to get

$$J = (-y_k + 2u_k)^2 + y_k^2 + u_k^2 = 2y_k^2 - 4y_k u_k + 5u_k^2$$

Note that $\arg \min_{u_k} 2y_k^2 - 4y_k u_k + 5u_k^2 = \arg \min_{u_k} -4y_k u_k + 5u_k^2$

Since $2y_k^2$ is independent of u_k

Hence we can instead solve

$$\min_{u_k} \underbrace{u_k^T 5}_{H} u_k - \underbrace{4y_k}_{h^T} u_k$$

The constraint:

$$|u_k| \leq 1 \Leftrightarrow -1 \leq u_k, \quad u_k \leq 1 \Leftrightarrow -u_k \leq 1, \quad u_k \leq 1$$

which can be expressed as

$$\underbrace{\begin{bmatrix} -1 \\ 1 \end{bmatrix}}_L u_k \leq \underbrace{\begin{bmatrix} 1 \\ 1 \end{bmatrix}}_b$$

Note: The comparison is element wise.

$$H = 5 \quad h^T = -4y_k \quad L = \begin{bmatrix} -1 \\ 1 \end{bmatrix} \quad b = \begin{bmatrix} 1 \\ 1 \end{bmatrix}$$